

Question 1

The averaged circuit model of an ideal buck converter is shown in Figure Q1. The buck converter has $L = 75 \mu\text{H}$, $C = 330 \mu\text{F}$, input voltage, $v_d = 18 \text{ V}$, output voltage, $v_o = 12 \text{ V}$, load resistance, $R = 6 \Omega$, and switching frequency $f_s = 100 \text{ kHz}$. The circuit operates in continuous conduction mode.

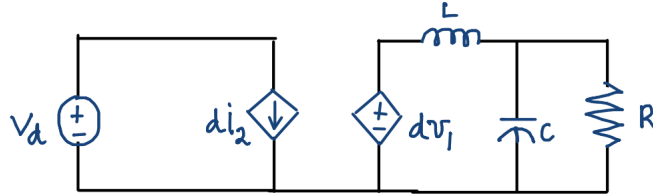


Figure Q1

- (i) By doing perturbation and linearization on the averaged circuit model, obtain AC small-signal model and DC steady-state model. [5 marks]
- (ii) Analyze DC steady-state model to determine all bias point constants. [5 marks]

Question 2

Analyze AC small signal model to obtain control to output transfer function, $\frac{\hat{v}_o}{\hat{d}}(s)$ for the AC small signal circuit in Figure Q2. A and B are circuit constants.

[3 marks]

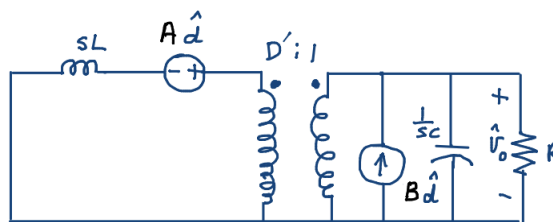


Figure Q2

Question 3

A boost converter has $L = 22 \mu\text{H}$, $C = 440 \mu\text{F}$, equivalent series resistance of capacitor, $R_{\text{ESR}} = 15 \text{ m}\Omega$, input voltage, $V_d = 12 \text{ V}$, output voltage, $v_o = 18 \text{ V}$, load resistance, $R = 6 \Omega$, and switching frequency $f_s = 100 \text{ kHz}$. The PWM has the sawtooth peak $V_m = 2 \text{ V}$ and the error amplifier has $V_{\text{ref}} = 2 \text{ V}$. The circuit operates in continuous conduction mode.

It is desired to design a type-3 compensated error amplifier, as shown in the **Appendix B**, based on k-factor approach, that will give a phase margin of at least 45° at the crossover frequency of 4 kHz .

- (i) Determine the magnitude in dB and phase in degrees of the loop gain at the crossover frequency, excluding error amplifier. [5 marks]
- (ii) What are the gain and the phase boost of the error amplifier required at the crossover frequency? [2 marks]
- (iii) Determine all resistance and capacitance values of the type-3 compensated error amplifier (see Appendix A). [4 marks]

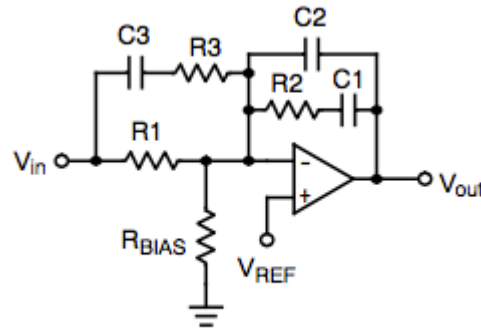
Potentially Useful Formula

| Transfer Function | With $s = j\omega$ | Gain | Phase |
|--|--|--|--|
| $1 + \frac{s}{\omega_z}$ | $1 + j\frac{\omega}{\omega_z}$ | $\sqrt{1 + \left(\frac{\omega}{\omega_z}\right)^2}$ | $\tan^{-1}\left(\frac{\omega}{\omega_z}\right)$ |
| $1 - \frac{s}{\omega_z}$ | $1 - j\frac{\omega}{\omega_z}$ | $\sqrt{1 + \left(\frac{\omega}{\omega_z}\right)^2}$ | $-\tan^{-1}\left(\frac{\omega}{\omega_z}\right)$ |
| $\frac{1}{1 + \frac{s}{\omega_p}}$ | $\frac{1}{1 + j\frac{\omega}{\omega_p}}$ | $\frac{1}{\sqrt{1 + \left(\frac{\omega}{\omega_p}\right)^2}}$ | $-\tan^{-1}\left(\frac{\omega}{\omega_p}\right)$ |
| $\frac{1}{1 + \frac{s}{Q\omega_o} + \left(\frac{\omega}{\omega_o}\right)^2}$ | $\frac{1}{1 - \left(\frac{\omega}{\omega_o}\right)^2 + j\frac{\omega}{Q\omega_o}}$ | $\frac{1}{\sqrt{\left[1 - \left(\frac{\omega}{\omega_o}\right)^2\right]^2 + \left(\frac{\omega}{Q\omega_o}\right)^2}}$ | $-\tan^{-1}\left(\frac{\omega/Q\omega_o}{1 - \left(\frac{\omega}{\omega_o}\right)^2}\right)$ |

| Converter | Control-to-output transfer function, $\frac{\hat{v}_o}{\hat{d}}$ | Parameters definition |
|------------|--|--|
| Buck | $g_{do} \frac{\left(1 + \frac{s}{\omega_{zESR}}\right)}{1 + \frac{s}{Q\omega_o} + \frac{s^2}{\omega_o^2}}$ | $g_{do} = V_d ; \omega_o = \frac{1}{\sqrt{LC}}$ $\omega_{zESR} = \frac{1}{rC} ; Q = \frac{R}{\omega_o L}$ |
| Boost | $g_{do} \frac{\left(1 - \frac{s}{\omega_{zRHP}}\right)\left(1 + \frac{s}{\omega_{zESR}}\right)}{1 + \frac{s}{Q\omega_o} + \frac{s^2}{\omega_o^2}}$ | $g_{do} = \frac{V_d}{(1-D)^2} ; \omega_o = \frac{1}{\sqrt{L_{eq}C}}$ $\omega_{zESR} = \frac{1}{rC} ; \omega_{zRHP} = \frac{R}{L_{eq}}$ $Q = \frac{R}{\omega_o L_{eq}} ; L_{eq} = \frac{L}{(1-D)^2}$ |
| Buck-boost | $g_{do} \frac{\left(1 - \frac{s}{\omega_{zRHP}}\right)\left(1 + \frac{s}{\omega_{zESR}}\right)}{1 + \frac{s}{Q\omega_o} + \frac{s^2}{\omega_o^2}}$ | $g_{do} = \frac{V_d}{(1-D)^2} ; \omega_o = \frac{1}{\sqrt{L_{eq}C}}$ $\omega_{zESR} = \frac{1}{rC} ; \omega_{zRHP} = \frac{R}{DL_{eq}}$ $Q = \frac{R}{\omega_o L_{eq}} ; L_{eq} = \frac{L}{(1-D)^2}$ |

Appendix B

K-factor approach for type 3 error amplifier



1. Calculate the gain, A_{co} (in dB), and the phase, P_{co} (in degree), of the power stage including PWM at the cross over frequency, f_{co} .
2. Calculate the amplifier phase boost, P_{boost} , and gain, G , required:

$$P_{boost} = PM - 90^\circ - P_{co}$$

$$G = 10^{\frac{-A_{co}}{20}}$$

3. Calculate K factor:

$$K = \tan^2 \left(\frac{P_{boost}}{4} + 45^\circ \right)$$

4. Calculate capacitances and resistances consecutively:

$$R1 = 4.7 \text{ k}\Omega$$

$$R_{bias} = \frac{V_{ref} R_1}{V_o - V_{ref}}$$

$$C2 = \frac{1}{2\pi f_{co} G R1}$$

$$C1 = C2(K - 1)$$

$$R2 = \frac{\sqrt{K}}{2\pi f_{co} C1}$$

$$R3 = \frac{R1}{(K - 1)}$$

$$C3 = \frac{1}{2\pi f_{co} \sqrt{K} R3}$$